

DAC14135 14-bit, 135MSPS D/A Converter

General Description

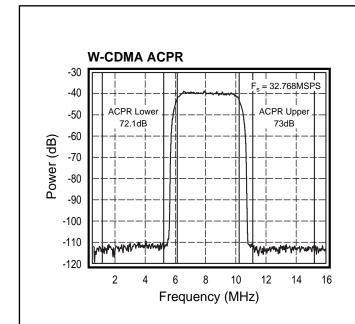
The DAC14135 is a monolithic 14-bit, 135MSPS digital-to-analog converter. The device has been optimized for use in cellular base stations and other applications where high resolution, high sampling rate, wide dynamic range, and compact size are required. The DAC14135 has many integrated features including a proprietary segmented DAC core, differential current outputs, a band-gap voltage reference, and TTL/CMOS compatible inputs. The converter features an 85dBc spurious free dynamic range (SFDR) at low frequencies and a 70dBc SFDR with 20MHz output signals. The 48-pin TSSOP package provides an extremely small footprint for applications where space is a critical consideration. The DAC14135 operates from a single +5V power supply. The digital power supply can also operate from +3.3V for lower power consumption and compatibility with +3.3V data inputs. The DAC14135 is fabricated in a 0.5µm CMOS process and is specified over the industrial temperature range of -40°C to +85°C. National Semiconductor thoroughly tests each part to verify full compliance with the guaranteed specifications.

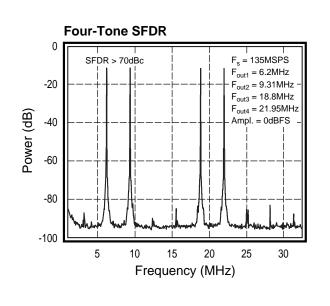
Features

- 135 MSPS
- Wide dynamic range SFDR @ 1MHz f_{out}: 85dBc SFDR @ 5MHz f_{out}: 79dBc SFDR @ 20MHz f_{out}: 70dBc
- Differential Current Outputs
- Low power consumption: 185mW
- Very small package: 48-pin TSSOP
- TTL/CMOS (+3.3V or +5V) inputs

Applications

- Cellular Basestations:
 GSM, WCDMA, DAMPS, etc.
- · Multi-carrier Basestations
- Multi-standard Basestations
- Direct digital synthesis (DDS)
- ADSL modems
- · HFC modems





PARAMETERS	CONDITIONS	TEMP		RATINGS	UNITS	NOTES	
			MIN	TYP	MAX		
RESOLUTION FULL SCALE CURRENT MAXIMUM CONVERSION RATE SFDR (1 ST Nyquist band) SFDR (1 ST Nyquist band) SFDR (1 ST Nyquist band) NOISE FLOOR	f_{out} = 1MHz, 0dBFS f_{out} = 5MHz, 0dBFS f_{out} = 20MHz, 0dBFS f_{out} = 5MHz, 0dBFS	Full Full Full Full Full +25°C	135 75 70 64	14 20 150 85 79 70 -146		Bits mA MSPS dBc dBc dBc dBFS/Hz	1 1,2 2 2 1,2
DYNAMIC LINEARITY @ DV _{DD} = +5V spurious-free dynamic range f _{out} = 1MHz f _{out} = 5MHz f _{out} = 20MHz SFDR within a band four-tone SFDR	sample rate = 135MSPS 1 ST Nyquist band 0dBFS 0dBFS 0dBFS fout = 20MHz, 4MHz band 6.2, 9.31, 18.8, 21.95 MHz	Full Full Full +25°C +25°C	75 70 64	85 79 70 90 72		dBc dBc dBc dBc dBc	2 2 1, 2
DYNAMIC LINEARITY @ DV _{DD} = +3.3V spurious-free dynamic range f _{out} = 1MHz f _{out} = 5MHz f _{out} = 20MHz	sample rate = 100MSPS 1^{ST} Nyquist band $0dBFS$, $DV_{DD} = +3.3V$ $0dBFS$, $DV_{DD} = +3.3V$ $0dBFS$, $DV_{DD} = +3.3V$	+25°C +25°C +25°C		83 77 70		dBc dBc dBc	
DYNAMIC CHARACTERISTICS glitch impulse settling time to 0.1% rise time fall time	step size = I _{fullscale} /2	+25°C +25°C +25°C +25°C		1 30 0.4 0.4		pV-s ns ns ns	3
DC ACCURACY AND PERFORMANCE differential non-linearity integral non-linearity gain error gain drift offset error reference voltage	20mA output current	+25°C +25°C +25°C Full +25°C +25°C	1.111	±1.0 ±1.5 ±5.0 ±75 10	1.358	LSB LSB % of FS ppm/°C nA V	
ANALOG OUTPUT PERFORMANCE full scale current compliance voltage (high) compliance voltage (low) output resistance output capacitance	at mid-scale at mid-scale	+25°C +25°C +25°C +25°C +25°C		20 1.25 -0.5 150 8.5		mA V V kΩ pF	
DATA INPUTS input logic low voltage, V _{IL} input logic high voltage, V _{IH} input logic low voltage, V _{IL} input logic high voltage, V _{IH} input logic high voltage, V _{IH} input logic low current, I _{IL} input logic high current, I _{IH}	DV _{DD} = +3.3V DV _{DD} = +3.3V	Full Full Full Full Full	3.5 2.4 -10 -10		1.3 0.9 10 10	> > >	1 1 1 1 1
TIMING maximum conversion rate setup time (T _S) hold time (T _H) propagation delay (T _{PD}) latency		Full +25°C +25°C +25°C +25°C	135 0.5 4.5	150 2 1		MSPS ns ns ns clk cycles	1, 2
CLOCK INPUTS clock inputs internal self bias differential clock input swing differential clock input slew rate clock input impedance (single-ended)		+25°C Full Full +25°C	1.5 1	1.5 1.2		V Vpp V/ns kΩ	

Min/max ratings are based on product characterization and simulation. Individual parameters are tested as noted. Outgoing quality levels are determined from tested parameters.

DAC14135 Electrical Characteristics

(sample rate = 135MSPS, T_{min} = -40°C, T_{max} = +85°C, AV_{DD} = +5V, DV_{DD} = +5V, CV_{DD} = +5V, full scale current = 20mA, differential 50 Ω doubly terminated output, unless specified otherwise)

PARAMETERS	CONDITIONS	TEMP	RATINGS			UNITS	NOTES
			MIN	TYP	MAX		
POWER REQUIREMENTS analog supply current digital supply current digital supply current power consumption power consumption AVDD power supply rejection ratio	135MSPS, $DV_{DD} = +5V$ 100MSPS, $DV_{DD} = +3.3V$ 135MSPS, $DV_{DD} = +5V$ 100MSPS, $DV_{DD} = +3.3V$ at mid-scale	+25°C +25°C +25°C +25°C +25°C +25°C		28 9 4.5 185 150 1.0	35 15	mA mA mA mW mW	1 1

Min/max ratings are based on product characterization and simulation. Individual parameters are tested as noted. Outgoing quality levels are determined from tested parameters.

Notes

- 1) These parameters are 100% tested at 25°C.
- 2) These parameters are sample tested at -40°C, +25°C and +85°C.
- Defined as the net area of undesired output transients in pV-s at a major transition.

Absolute Maximum Ratings

Note: Absolute maximum ratings are limiting values, to be applied individually, and beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied. Exposure to maximum ratings for extended periods may affect device reliability.

Recommended Operating Conditions

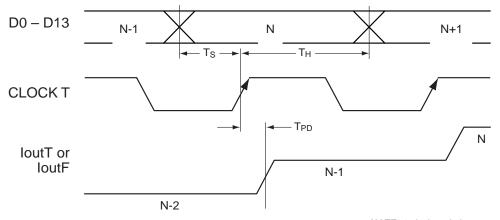
positive analog supply voltage $+5V \pm 5\%$ positive digital supply voltage $+3.3V \text{ or } +5V \pm 5\%$ positive clock supply voltage $+5V \pm 5\%$ operating temperature range +60% C to +85% C

Package Thermal Resistance								
Package	$\theta_{\sf JA}$	θЈС						
48-pin TSSOP	56°C/W	16°C/W						

Package Transistor Count

Transistor count 8,600

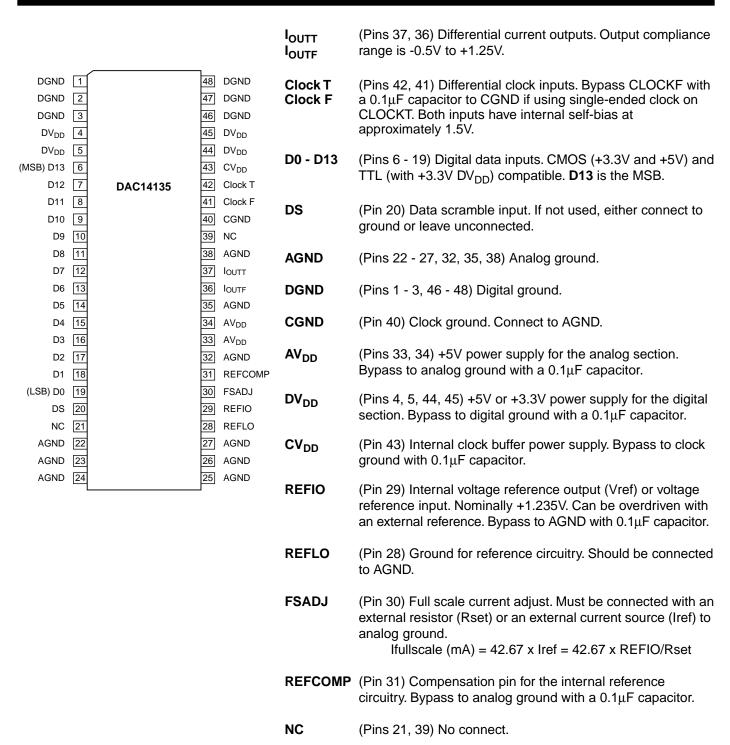
Ordering Information										
Model	Temperature Range	Description								
DAC14135MT DAC14135MTX DAC14135PCASM	-40°C to +85°C -40°C to +85°C	48-pin TSSOP (industrial temperature range) 48-pin TSSOP (TNR 1000 pc reel) Fully loaded evaluation board with DAC14135 ready for test.								



NOTE: 1 clock cycle latency

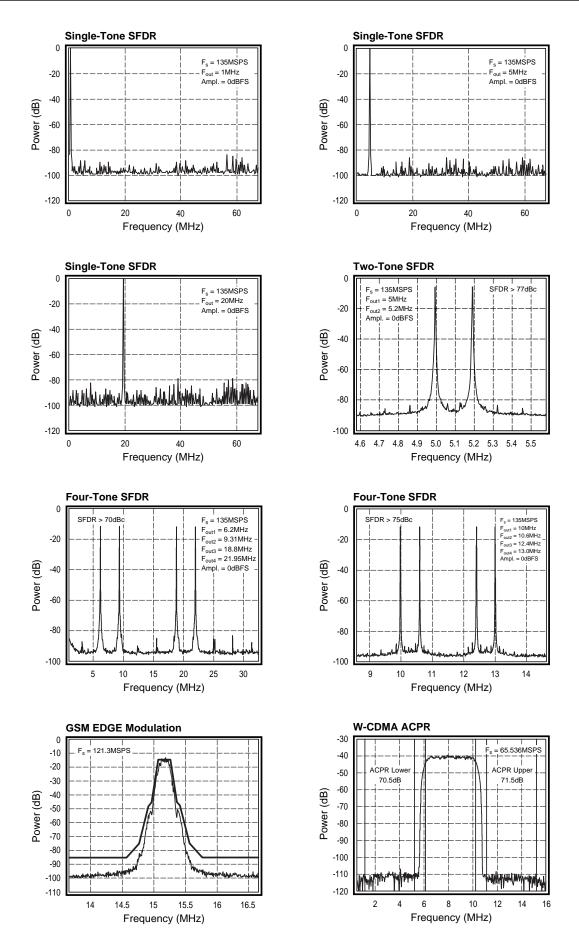
DAC14135 Timing Diagram

DAC14135 Pin Definitions

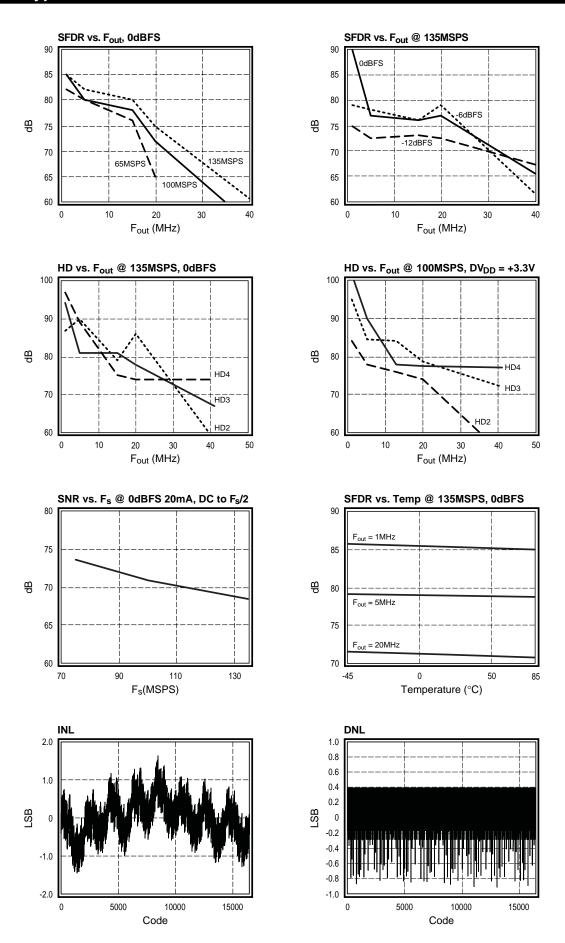


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DAC14135 Typical Performance Characteristics (AV_{DD}= +5V, DV_{DD} = +5V, CV_{DD} = +5V, T_A = 25°C)



$\textbf{DAC14135 Typical Performance Characteristics} \ (AV_{DD} = +5V, \, DV_{DD} = +5V, \, CV_{DD} = +5V, \, T_A = 25^{\circ}C)$



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DAC14135 Application Information

Digital Data Inputs

The DAC14135's 14-bit binary inputs are CMOS compatible. The input voltage thresholds are approximately half of the digital supply voltage (DV_{DD}/2). For a 3.3V DV_{DD}, the inputs are also compatible with standard TTL levels. Digital data is standard binary coded, D13 is the most significant bit and D0 is the least significant bit. For all 1's at the input, $I_{OUTT} = I_{fullscale}$, $I_{OUTF} = 0$. For all 0's at the input, $I_{OUTT} = 0$, $I_{OUTF} = I_{fullscale}$.

To prevent or reduce digital data feedthrough, keep digital data lines short and ensure separate digital grounding (DGND). 75Ω resistors in series with the digital data input path may be used to reduce overshoot and data feedthrough to the analog outputs. Digital supply (DV_{DD}) should be decoupled to DGND using a $0.1\mu F$ bypass capacitor.

Driving the Clock Inputs

The differential clock inputs, Clock T and Clock F, may be driven by a variety of input sources. These pins are internally self-biased at about 1.5V and therefore can be differentially AC coupled. Alternatively, a single clock source on Clock T with Clock F bypassed to CGND using a 0.1μF capacitor, may be used to clock the DAC14135. The clock driver supply voltage (CVDD) should be 5V ±5% and should be decoupled to the clock ground (CGND) using a 0.1μF capacitor. For best SFDR performance, use a differential clock input. Minimum input voltage swing $(1.5V_{pp})$ and slew rate (1.0V/ns)requirements should be met for optimum performance. Low noise and low jitter clocks provide the best SNR performance for the DAC14135. Figure 1 shows one method of driving the clock inputs. A low noise sinusoidal clock source (2-4 V_{DD}) may be used to drive the transformer primary.

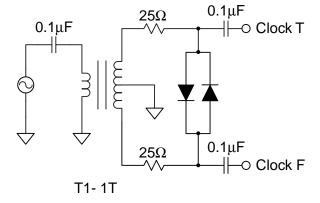


Figure 1: Method of Driving Clock Inputs

The transformer converts the single ended clock signal to a differential signal. The diodes in the secondary limit the input swing to the DAC14135.

Latching the Input Data

Inputs of the DAC14135 include a master-slave flip-flop. Due to internal clock buffer delay, the DAC14135 requires more hold time than setup time. This timing should be observed at the DAC data and clock pins. Refer to the timing diagram and the specifications for proper setup and hold time requirements.

Data Scramble (DS) Input Pin

The DAC14135 is equipped with a data scramble input pin (DS) that may be used to troubleshoot possible spurious or harmonic distortion degradation due to digital data feedthrough on the printed circuit board. In the DAC14135, the digital data inputs are logically XORed with the DS input pin as shown in Figure 2.

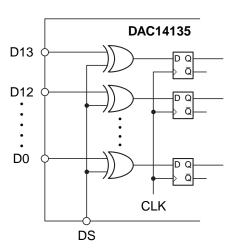


Figure 2: Digital Data Inputs with DS Input Pin

If the DS pin is at logic low (DGND) the input data is left unchanged and if this pin is at logic high (DV_{DD}) the input data is inverted. If the input data is XORed with a random bit stream and if the same random bit stream is used to drive the DS pin, low order harmonics due to data feedthrough on the printed circuit board can be reduced. If this feature is not used, tie DS pin to ground or leave it floating (DS pin has internal active pulldown).

Voltage Reference Loop

The DAC14135 has an internal bandgap voltage reference nominally at 1.235V. The output of this bandgap is connected to the REFIO pin. The REFIO pin is a high impedance output and therefore can be easily over-

ridden by an external bandgap reference voltage. The reference ground (REFLO) should always be tied to analog ground. The REFIO pin should be bypassed to REFLO using a $0.1\mu F$ capacitor. For reduced noise, an external compensation capacitor (0.1 μF) should also be used to bypass the internal reference loop from pin REFCOMP to AGND. Figure 3 shows the internal voltage reference loop functional schematic.

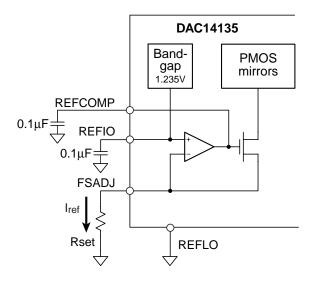


Figure 3: Internal Voltage Loop
Functional Schematic

A reference current source (Iref) from pin FSADJ to ground may be used to set the full scale output current (Ifs) of the DAC14135. The full scale current is given by,

Ifs =
$$42.67 \times Iref$$

Alternatively, a resistor (Rset) from FSADJ to AGND may be used to set the full scale output current of the DAC.

Ifs
$$(mA) = 42.67 \times REFIO/Rset$$

The voltage at REFIO is nominally set by the internal bandgap at 1.235V. For a full scale output current of 20mA, the value of Rset is $2.635 k\Omega$.

Analog Outputs

The differential analog outputs, I_{OUTT} and I_{OUTF} are high impedance current source outputs. These outputs, if terminated into 50Ω at 20mA full scale current, will generate a differential voltage output at $2V_{pp}$. The output compliance of each of the current outputs of the DAC14135 is -0.5V to +1.25V. The differential outputs can be converted to a single-ended output using an RF center-tapped transformer or a differential to single-ended amplifier. The I_{OUTT} and I_{OUTF} traces on the printed circuit board should be short and matched with adequate analog grounding nearby. One example of an AC coupled differential to single-ended topology is shown in Figure 4.

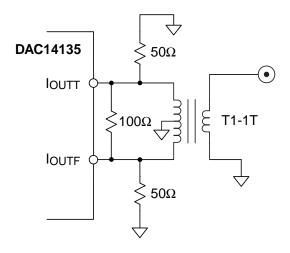


Figure 4: AC Coupled Differential to Single-ended Topology

DAC14135 Grounding Information

In the DAC14135, all the grounds AGND, REFLO, DGND and CGND are shorted together inside the package. The purpose of having separate grounds on the printed circuit board is to prevent digital data currents from returning through the analog or reference grounds, and corrupting the analog outputs. Refer to the evaluation board layout.

DAC14135 Evaluation Board Description

General Description

The DAC14135 Evaluation Board is intended to aid in evaluating the performance of the DAC14135. The board allows the user to exercise the inputs to the DAC and examine the output in either differential or single ended mode. The board comes complete with the DAC14135, a transformer network to convert a single ended clock to a differential clock, a transformer to convert the differential output from I_{OUTT} and I_{OUTF} to a single ended output, and an edge connector. This is a 5V part, but if a 3.3V CMOS or TTL digital data interface is required, the digital supply (DV_{DD}) should be 3.3V. A 3.3V regulator is provided so that the board can be run off of a single 5V supply. For the best distortion performance at the maximum clock frequency, D_{VDD} should be set to 5V.

Setup and Configuration

There are two terminal blocks on the DAC14135 evaluation board, one in the upper left corner next to the AMP connectors, and one in the upper right corner. The upper right corner has the analog power supply connector, marked $+A_{VDD}$. The connector in the upper left is for the digital power supply and is marked $+D_{VDD}$. There is also a jumper next to the $+D_{VDD}$ terminal block marked D_{VDD} with one end marked DIRECT and the other end marked +3.3V REG.

There are three ways to power the evaluation board. The default method of use is to connect the 5V power supply to both the $+A_{VDD}$ terminal block and the $+D_{VDD}$ terminal block and connect the jumper between the DIRECT pin (pin 1) and the middle pin (pin 2).

If a 3.3V CMOS or TTL digital data interface is required, connect the jumper between the +3.3V REG pin (pin 3) and the middle pin (pin 2). This enables the 3.3V regulator on the back side of the board. The output of the regulator is filtered and powers the digital portion of the DAC.

To use the board in the dual supply mode, connect a 5V supply to the $+A_{VDD}$ terminal block, connect a 3.3V supply to the $+D_{VDD}$ terminal block and connect the jumper between the DIRECT pin (pin 1) and the middle pin (pin 2). This bypasses the on-board voltage regulator, although the regulator still draws power.

Getting Data to the Evaluation Board

The DAC14135 evaluation board is shipped with the edge connectors J1 and J2 being the default data input interface. J1 and J2 are AMP 536511-1 and 536511-3 edge connectors respectively. Data should be at the same voltage level as D_{VDD} . Figure 5 below, is an edge-on view of J2. Pins 24D-11D are the data lines with 24D being the MSB. The ground pins are 23C, 23A, 21C, 19C, 17C, 17A, 15C, 13C, 11C, 11A, 9C, 7C, 6A, 5C, 3C, and 1C. All ground pins are tied together on-board. Also, pin 10D should be at logic LOW (0V) if the data scramble feature on the DAC14135 is not used.

Driving the Clock Input

The evaluation board has an on-board transformer, T2, that converts a single ended clock to a differential clock to drive the DAC14135. For best results drive the CLOCK SMA connector with a low jitter 50Ω source. If a sinusoidal source is used, its peak-to-peak amplitude should be at least 2.5V to meet the minimum clock input slew rate requirement. Back-to-back diodes at the secondary of the transformer T2 limit the voltage swing at the DAC14135 Clock T and Clock F input pins.

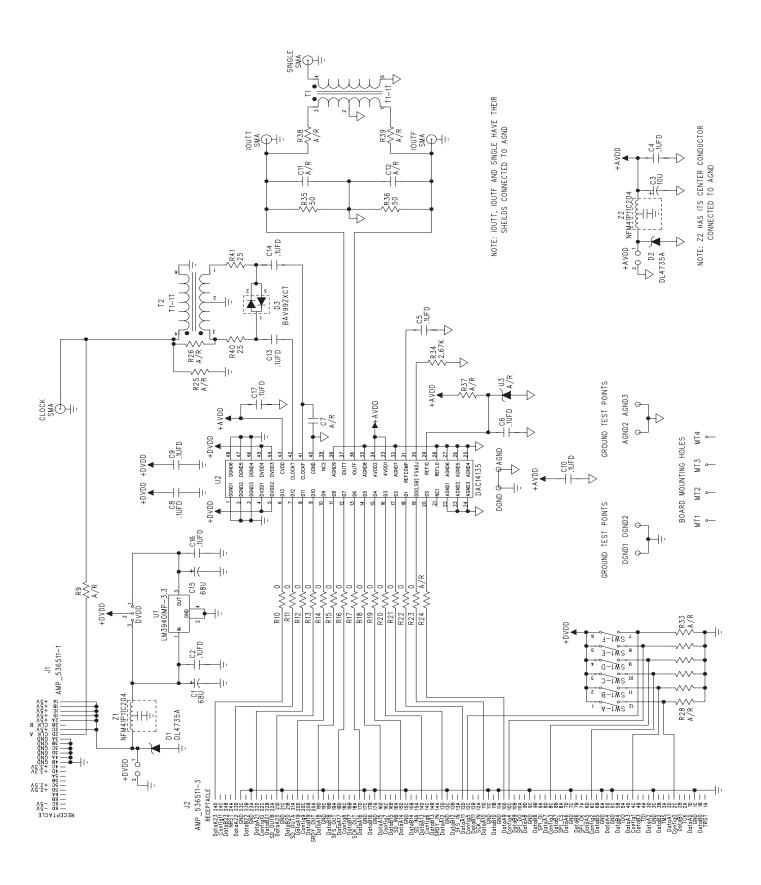
Measuring the Analog Outputs

The evaluation board is shipped with transformer T1 installed to convert the differential output to a single ended output. However, the 0Ω resistors R38 and R39 are not installed. To take single ended measurements, install R38 and R39 and attach your instrument to the SMA connector marked 'SINGLE'. For differential output measurements, remove R38 and R39 if they are installed. Note that both outputs, I_{OUTT} and I_{OUTF} are terminated with 50Ω .

24D	23D	22D	21D	20D	19D	18D	17D	16D	15D	14D	13D	12D	11D	10D	9D	8D	7D	6D	5D	4D	3D	2D	1D
24C	23C	22C	21C	20C	19C	18C	17C	16C	15C	14C	13C	12C	11C	10C	9C	8C	7C	6C	5C	4C	3C	2C	1C
24B	23B	22B	21B	20B	19B	18B	17B	16B	15B	14B	13B	12B	11B	10B	9B	8B	7B	6B	5B	4B	3B	2B	1B
24A	23A	22A	21A	20A	19A	18A	17A	16A	15A	14A	13A	12A	11A	10A	9A	8A	7A	6A	5A	4A	3A	2A	1A

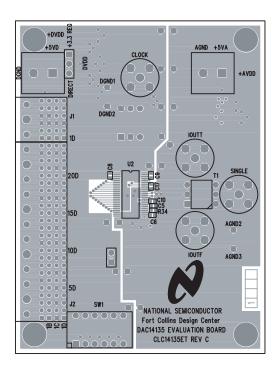
Figure 5: Pinout for J2 (Amp 536511-3)

DAC14135 Evaluation Board Schematic

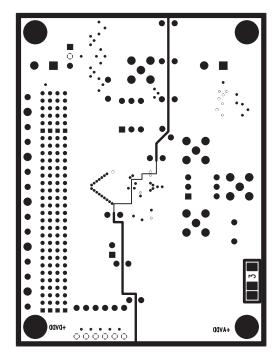


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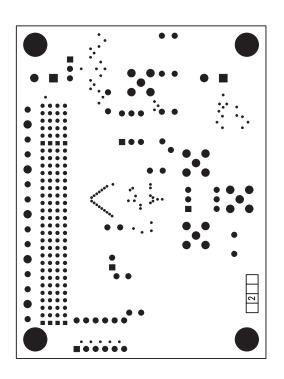
DAC14135 Evaluation Board Layout



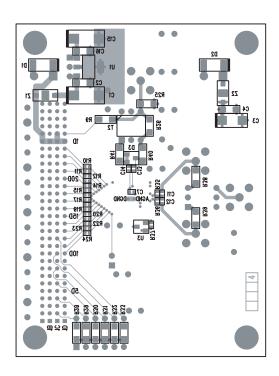
DAC14135PCASM Layer 1



DAC14135PCASM Layer 3

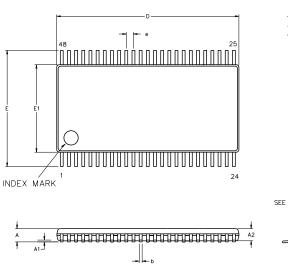


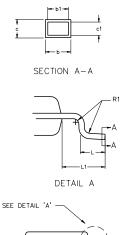
DAC14135PCASM Layer 2



DAC14135PCASM Layer 4

DAC14135 Physical Dimensions





Symbol	Min Max		Notes
Α	_	1.10	
A1	0.05	0.15	
A2	0.80	1.05	
b	0.17	0.27	
b1	0.17	0.23	
С	0.09	0.20	
c1	0.09	0.16	
D	12.40	12.60	2
E	8.1		
E1	6.00	6.20	2
е	0.50		
L	0.50	0.75	
L1	1.00		
R1	0.127	0.127	

Notes:

- 1. All dimensions are in millimeters.
- 2. Dimensions D and E1 do not include mold protrusion. Allowable protrusion is 0.20mm per side.

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